



AN INTRODUCTION TO SYSTEM-THEORETIC METHODS FOR MODEL REDUCTION Part V: Balancing-based Methods for Nonlinear Systems Peter Benner Pawan K. Goyal Igor Pontes Duff obias Breiten (KFU Graz \rightarrow TU Berlin) Tobias Damm (TU Kaiserslautern) **Special Semester on** and dimension reduction in uncertain and dynamic systems" ICERM at Brown University, Providence, RI, February 13, 2020

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- 1. Introduction
- 2. Gramian-based Model Reduction for Linear Systems
- 3. Balanced Truncation for Bilinear Systems
- 4. Balanced Truncation for QB Systems
- 5. Balanced Truncation for Polynomial Systems



1. Introduction

Model Reduction for Control Systems System Classes How general are these system classes? Linear Systems and their Transfer Functions

2. Gramian-based Model Reduction for Linear Systems

- 3. Balanced Truncation for Bilinear Systems
- 4. Balanced Truncation for QB Systems
- 5. Balanced Truncation for Polynomial Systems



Introduction Model Reduction for Control Systems

Nonlinear Control Systems

$$\Sigma : \begin{cases} E\dot{x}(t) = f(t, x(t), u(t)), & Ex(t_0) = Ex_0, \\ y(t) = g(t, x(t), u(t)), \end{cases}$$

with

- (generalized) states $x(t) \in \mathbb{R}^n$,
- inputs $u(t) \in \mathbb{R}^m$,
- outputs $y(t) \in \mathbb{R}^q$.

If *E* singular \rightsquigarrow descriptor system. Here, $E = I_n$ for simplicity.





Original System ($E = I_n$)

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Goals:

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u||$ for all admissible input signals.



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Reduced-Order Model (ROM)

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 - states $\hat{x}(t) \in \mathbb{R}^r$, $r \ll n$,
 - inputs $u(t) \in \mathbb{R}^m$,
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\xrightarrow{u} Σ \xrightarrow{y}

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Goals:

 $||y - \hat{y}|| < \text{tolerance} \cdot ||u||$ for all admissible input signals. Secondary goal: reconstruct approximation of x from \hat{x} .



$$\begin{split} \dot{x}(t) &= f(t,x,u) = \mathcal{A}(x(t)) + \mathcal{B}(x(t))u(t), \quad \mathcal{A}: \mathbb{R}^n \to \mathbb{R}^n, \ \mathcal{B}: \mathbb{R}^n \to \mathbb{R}^{n \times m}, \\ y(t) &= g(t,x,u) = \mathcal{C}(x(t)) + \mathcal{D}(x(t))u(t), \quad \mathcal{C}: \mathbb{R}^n \to \mathbb{R}^q, \ \mathcal{D}: \mathbb{R}^n \to \mathbb{R}^{q \times m}. \end{split}$$



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Linear, Time-Invariant (LTI) Systems

$$\begin{aligned} \dot{x}(t) &= f(t,x,u) &= Ax(t) + Bu(t), & A \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \\ y(t) &= g(t,x,u) &= Cx(t) + Du(t), & C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}. \end{aligned}$$



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Bilinear Systems

 $\begin{aligned} \dot{x}(t) &= f(t, x, u) = Ax(t) + \sum_{i=1}^{m} u_i(t)A_ix(t) + Bu(t), \quad A, A_i \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \\ y(t) &= g(t, x, u) = Cx(t) + Du(t), \qquad \qquad C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}. \end{aligned}$



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Quadratic-Bilinear (QB) Systems

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Polynomial Systems

$$\dot{x}(t) = f(t, x, u) = Ax(t) + \sum_{j=2}^{n_p} H_j\left(\otimes^j x(t)\right) + \sum_{j=2}^{n_p} \sum_{k=1}^m A_j^k\left(\otimes^j x(t)\right) u_k(t) + Bu(t),$$

$$H_j, A_j^k \text{ of "appropriate size"},$$

$$y(t) = g(t, x, u) = Cx(t) + Du(t), \qquad C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}.$$



$$\begin{split} \dot{x}(t) &= f(t,x,u) = \mathcal{A}(x(t)) + \mathcal{B}(x(t))u(t), \quad \mathcal{A}: \mathbb{R}^n \to \mathbb{R}^n, \ \mathcal{B}: \mathbb{R}^n \to \mathbb{R}^{n \times m}, \\ y(t) &= g(t,x,u) = \mathcal{C}(x(t)) + \mathcal{D}(x(t))u(t), \quad \mathcal{C}: \mathbb{R}^n \to \mathbb{R}^q, \ \mathcal{D}: \mathbb{R}^n \to \mathbb{R}^{q \times m}. \end{split}$$

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Written in control-affine form:

$$\begin{aligned} \mathcal{A}(x) &:= Ax + H(x \otimes x), \qquad \mathcal{B}(x) &:= [A_1, \dots, A_m](I_m \otimes x) + B \\ \mathcal{C}(x) &:= Cx, \qquad \qquad \mathcal{D}(x) &:= D. \end{aligned}$$



$$\dot{x} = \mathcal{A}(x) + Bu$$
 with $\mathcal{A}(0) = 0$,
 $y = Cx + Du$.



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Instead of truncating Taylor expansion, Carleman (bi)linearization takes into account K higher order terms (h.o.t.) by introducing new variables:

$$x^{(k)} := x \underbrace{\otimes \cdots \otimes}_{(k-1) \text{ times}} x, \qquad k = 1, \dots, K.$$

Here: K = 2, i.e., $z := x^{(2)} = x \otimes x$.



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Instead of truncating Taylor expansion, Carleman (bi)linearization takes into account K = 2 higher order terms (h.o.t.) by introducing new variables: $z := x^{(2)} = x \otimes x$. Then z satisfies

$$\dot{z} = \dot{x} \otimes x + x \otimes \dot{x} = (Ax + Hz + \ldots + Bu) \otimes x + x \otimes (Ax + Hz + \ldots + Bu).$$



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Ignoring h.o.t. \implies bilinear system with state $x^{\otimes} := [x^T, z^T]^T \in \mathbb{R}^{n+n^2}$:

$$\frac{d}{dt} x^{\otimes} = \begin{bmatrix} A & H \\ 0 & A \otimes I_n + I_n \otimes A \end{bmatrix} x^{\otimes} + \begin{bmatrix} 0 & 0 \\ B \otimes I_n + I_n \otimes B & 0 \end{bmatrix} (x^{\otimes}) u + \begin{bmatrix} B \\ 0 \end{bmatrix} u,$$

$$y^{\otimes} = \begin{bmatrix} C & 0 \end{bmatrix} x^{\otimes} + Du.$$



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Remark

Bilinear systems directly occur, e.g., in biological systems, PDE control problems with mixed boundary conditions, "control via coefficients", networked control systems, ...



QB systems can be obtained as approximation (by truncating Taylor expansion) to weakly nonlinear systems [PHILLIPS '03].

C. Gu. QLMOR: A Projection-Based Nonlinear Model Order Reduction Approach Using Quadratic-Linear Representation of Nonlinear Systems. IEEE TRANSACTIONS ON COMPUTER-AIDED DESIGN OF INTEGRATED CIRCUITS AND SYSTEMS, 30(9):1307–1320, 2011.

L. Feng, X. Zeng, C. Chiang, D. Zhou, and Q. Fang. Direct nonlinear order reduction with variational analysis. In: Proceedings of DATE 2004, pp. 1316-1321.

J. R. Phillips. Projection-based approaches for model reduction of weakly nonlinear time-varying systems. IEEE TRANSACTIONS ON COMPUTER-AIDED DESIGN OF INTEGRATED CIRCUITS AND SYSTEMS, 22(2):171-187, 2003.



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But exact representation of smooth nonlinear systems possible:

Theorem [Gu '09/'11]

Assume that the state equation of a nonlinear system is given by

 $\dot{x} = a_0 x + a_1 g_1(x) + \ldots + a_k g_k(x) + Bu,$

where $g_i(x) : \mathbb{R}^n \to \mathbb{R}^n$ are compositions of uni-variable rational, exponential, logarithmic, trigonometric or root functions, respectively. Then, by iteratively taking derivatives and adding algebraic equations, respectively, the nonlinear system can be transformed into a QB(DAE) system.

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Alternatively, polynomial-bilinear system can be obtained using iterated Lie brackets [Gu '11].

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FitzHugh-Nagumo model

Sine-Gordon equation

- Model describes activation and de-activation of neurons.
- Contains a cubic nonlinearity, which can be transformed to QB form.
- Applications in biomedical studies, mechanical transmission lines, etc.
- Contains sin function, which can also be rewritten into QB form.



Linear Systems in Frequency Domain

Application of Laplace transform $(x(t) \mapsto x(s), \dot{x}(t) \mapsto sx(s) - x(0))$ to linear system

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

with x(0) = 0 yields:

$$sx(s) = Ax(s) + Bu(s), \quad y(s) = Cx(s) + Du(s),$$

Model reduction in frequency domain: Fast evaluation of mapping $u \rightarrow y$.



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 \implies I/O-relation in frequency domain:

$$y(s) = \left(\underbrace{C(sI_n - A)^{-1}B + D}_{=:G(s)}\right)u(s).$$

G(s) is the **transfer function** of Σ .

Model reduction in frequency domain: Fast evaluation of mapping $u \rightarrow y$.



Formulating model reduction in frequency domain

Approximate the dynamical system

$$\begin{aligned} \dot{x} &= Ax + Bu, \qquad A \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \\ y &= Cx + Du, \qquad C \in \mathbb{R}^{q \times n}, \ D \in \mathbb{R}^{q \times m}, \end{aligned}$$

by reduced-order system

$$\begin{aligned} \dot{\hat{x}} &= \hat{A}\hat{x} + \hat{B}u, \quad \hat{A} \in \mathbb{R}^{r \times r}, \ \hat{B} \in \mathbb{R}^{r \times m}, \\ \hat{y} &= \hat{C}\hat{x} + \hat{D}u, \quad \hat{C} \in \mathbb{R}^{q \times r}, \ \hat{D} \in \mathbb{R}^{q \times m} \end{aligned}$$

of order $r \ll n$, such that

$$\|y - \hat{y}\| = \left\| \mathsf{G}u - \hat{\mathsf{G}}u \right\| \le \left\| \mathsf{G} - \hat{\mathsf{G}} \right\| \cdot \|u\| < \mathsf{tolerance} \cdot \|u\|$$
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Formulating model reduction in frequency domain

Approximate the dynamical system

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$$\|y - \hat{y}\| = \left\|Gu - \hat{G}u\right\| \le \left\|G - \hat{G}\right\| \cdot \|u\| < \text{tolerance} \cdot \|u\|.$$
oximation problem:
$$\min_{\text{order}(\hat{G}) \le r} \left\|G - \hat{G}\right\|.$$

 $\Rightarrow \mathsf{Appr}$



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/

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with A stable, i.e., $\Lambda(A) \subset \mathbb{C}^-$,
is balanced, if system Gramians, i.e., solutions P, Q of the Lyapunov equations
 $AP + PA^T + BB^T = 0, \qquad A^TQ + QA + C^TC = 0,$
satisfy: $P = Q = \operatorname{diag}(\sigma_1, \dots, \sigma_n)$ with $\sigma_1 \ge \sigma_2 \ge \cdots \ge \sigma_n \ge 0$



1



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- $\{\sigma_1, \ldots, \sigma_n\}$ are the Hankel singular values (HSVs) of Σ .
- Compute balanced realization (needs *P*, *Q*!) of the system via state-space transformation

$$\begin{aligned} \mathcal{T}: (A, B, C) &\mapsto (TAT^{-1}, TB, CT^{-1}) \\ &= \left(\left[\begin{array}{cc} A_{11} & A_{12} \\ A_{21} & A_{22} \end{array} \right], \left[\begin{array}{cc} B_1 \\ B_2 \end{array} \right], \left[\begin{array}{cc} C_1 & C_2 \end{array} \right] \right). \end{aligned}$$



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• Truncation $\rightsquigarrow (\hat{A}, \hat{B}, \hat{C}) = (A_{11}, B_1, C_1).$



Motivation:

HSV are system invariants: they are preserved under \mathcal{T} and determine the energy transfer given by the Hankel map

$$\mathcal{H}: L_2(-\infty, 0) \mapsto L_2(0, \infty): u_- \mapsto y_+.$$

"functional analyst's point of view"



Motivation:

HSV are system invariants: they are preserved under ${\cal T}$ and determine the energy transfer given by the Hankel map

$$\mathcal{H}: L_2(-\infty, 0) \mapsto L_2(0, \infty): u_- \mapsto y_+.$$

"functional analyst's point of view"

Minimum energy to reach x_0 in balanced coordinates:

$$\inf_{\substack{u \in L_2(-\infty,0]\\ x(0) = x_0}} \int_{-\infty}^0 u(t)^T u(t) \, dt = x_0^T P^{-1} x_0 = \sum_{j=1}^n \frac{1}{\sigma_j} x_{0,j}^2$$



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Energy contained in the system if $x(0) = x_0$ and $u(t) \equiv 0$ in balanced coordinates:

$$\|y\|_{2}^{2} = \int_{0}^{\infty} y(t)^{T} y(t) dt = x_{0}^{T} Q x_{0} = \sum_{j=1}^{n} \sigma_{j} x_{0,j}^{2}$$



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"engineer's point of view"

 \implies Truncate states corresponding to "small" HSVs



Recap: Balanced Truncation for Linear Systems

Properties

• Reduced-order model is stable with HSVs $\sigma_1, \ldots, \sigma_r$.



Properties

- Reduced-order model is stable with HSVs $\sigma_1, \ldots, \sigma_r$.
- Adaptive choice of r via computable error bound:

$$\|y - \hat{y}\|_{2} \leq \|G - \hat{G}\|_{\mathcal{H}_{\infty}} \|u\|_{2} \leq \left(2\sum_{k=r+1}^{n} \sigma_{k}\right) \|u\|_{2}.$$



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Practical implementation

- Rather than solving Lyapunov equations for P, Q (n^2 unknowns!), find $S, R \in \mathbb{R}^{n \times s}$ with $s \ll n$ such that $P \approx SS^T$, $Q \approx RR^T$.
- Reduced-order model directly obtained via small-scale $(s \times s)$ SVD of $R^T S!$
- No $\mathcal{O}(n^3)$ or $\mathcal{O}(n^2)$ computations necessary!

The concept of balanced truncation can be generalized to the class of bilinear systems, where we need the solutions of the Lyapunov-plus-positive equations:

$$AP + PA^{T} + \sum_{i=1}^{m} A_{i}PA_{i}^{T} + BB^{T} = 0,$$

 $A^{T}Q + QA^{T} + \sum_{i=1}^{m} A_{i}^{T}QA_{i} + C^{T}C = 0.$

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If unique solutions P = P^T ≥ 0, Q = Q^T ≥ 0 exist, these can be used in balancing procedure like for linear systems, with

 $\hat{A} := W^{\mathsf{T}} A V, \quad \hat{A}_i = W^{\mathsf{T}} A_i V, \quad \hat{B} := W^{\mathsf{T}} B, \quad \hat{C} := C V.$

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See [AL-BAIYAT/BETTAYEB 1993, B./DAMM 2011] for details.

• Stability preservation [B./DAMM/REDMANN/RODRIGUEZ CRUZ 2016].

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- These equations also appear for stochastic control systems, see [B./DAMM 2011].
- "Twice-the-trail-of-the-HSVs" error bound does not hold [B./DAMM 2014].
- Alternative Gramians based on linear matrix inequalities investigated by [REDMANN 2018], yield H_{∞} error bound based on truncated characteristic values, but hard to compute for large-scale systems!



Lyapunov-plus-Positive Equations

Some basic facts and assumptions

$$AX + XA^T + \sum_{i=1}^m A_i XA_i^T + BB^T = 0.$$

• Need a positive semi-definite symmetric solution X.

(1)



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- In standard Lyapunov case, existence and uniqueness guaranteed if A stable $(\Lambda(A) \subset \mathbb{C}^-)$; this is not sufficient here: (1) is equivalent to

$$\left(I_n\otimes A+A\otimes I_n+\sum_{i=1}^mA_i\otimes A_i\right)\operatorname{vec}(X)=-\operatorname{vec}(BB^{ op}).$$

Sufficient condition for unique solvability: smallness of A_i (related to stability radius of A). \rightsquigarrow **bounded-input bounded-output (BIBO) stability** of bilinear systems. This will be assumed from here on, hence: existence and uniqueness of positive semi-definite solution $X = X^T$.



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- Want: solution methods for large scale problems, i.e., only matrix-matrix multiplication with *A*, *A_j*, solves with (shifted) *A* allowed!
- Requires to compute data-sparse approximation to generally dense X; here: $X \approx ZZ^T$ with $Z \in \mathbb{R}^{n \times n_Z}$, $n_Z \ll n!$



Lyapunov-plus-Positive Equations Low-rank Approximations

Question

Can we expect low-rank approximations $ZZ^T \approx X$ to the solution of

$$AX + XA^{T} + \sum_{j=1}^{m} A_j XA_j^{T} + BB^{T} = 0 ?$$



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Standard Lyapunov case:

[GRASEDYCK '04]

$$AX + XA^T + BB^T = 0 \iff \underbrace{(I_n \otimes A + A \otimes I_n)}_{=:\mathcal{A}} \operatorname{vec}(X) = -\operatorname{vec}(BB^T).$$



[GRASEDYCK '04]

$$AX + XA^{T} + BB^{T} = 0 \iff \underbrace{(I_{n} \otimes A + A \otimes I_{n})}_{=:\mathcal{A}} \operatorname{vec}(X) = -\operatorname{vec}(BB^{T}).$$

Apply

$$M^{-1} = -\int_0^\infty \exp(tM) \mathrm{d}t$$

to ${\cal A}$ and approximate the integral via (sinc) quadrature \Rightarrow

$$\mathcal{A}^{-1} \approx -\sum_{i=-k}^{k} \omega_i \exp(t_k \mathcal{A}),$$

with error $\sim \exp(-\sqrt{k})$ ($\exp(-k)$ if $A = A^T$), then an approximate Lyapunov solution is given by

$$\operatorname{vec}(X) \approx \operatorname{vec}(X_k) = \sum_{i=-k}^k \omega_i \exp(t_i \mathcal{A}) \operatorname{vec}(BB^T).$$



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$$AX + XA^{T} + BB^{T} = 0 \iff \underbrace{(I_n \otimes A + A \otimes I_n)}_{=:A} \operatorname{vec}(X) = -\operatorname{vec}(BB^{T}).$$

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Now observe that

$$\exp(t_i\mathcal{A}) = \exp(t_i(I_n \otimes A + A \otimes I_n)) \equiv \exp(t_i\mathcal{A}) \otimes \exp(t_i\mathcal{A}).$$



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$$\operatorname{vec}(X_k) = \sum_{i=-k}^{k} \omega_i \left(\exp(t_i A) \otimes \exp(t_i A) \right) \operatorname{vec}(BB^T)$$
$$\implies X_k = \sum_{i=-k}^{k} \omega_i \exp(t_i A) B^T \exp(t_i A^T) \equiv \sum_{i=-k}^{k} \omega_i B_i B_i^T,$$

so that $rank(X_k) \leq (2k+1)m$ with

 $\|X - X_k\|_2 \lesssim \exp(-\sqrt{k})$ ($\exp(-k)$ for $A = A^T$)!



Lyapunov-plus-Positive Equations

Low-rank Approximations

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Problem: in general,

$$\exp\left(t_i(I\otimes A + A\otimes I + \sum_{j=1}^m A_j\otimes A_j)\right) \neq (\exp(t_iA)\otimes \exp(t_iA))\exp\left(t_i(\sum_{j=1}^m A_j\otimes A_j)\right).$$



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Assume that m = 1 and $A_1 = UV^T$ with $U, V \in \mathbb{R}^{n imes r}$ and consider

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Sherman-Morrison-Woodbury \Longrightarrow

$$\mathcal{A}\operatorname{vec}(X) = y + (U \otimes U) \underbrace{\left(I_r \otimes I_r - (V \otimes V)^T \mathcal{A}^{-1} (U \otimes U)\right) (V \otimes V)^T \mathcal{A}^{-1} y}_{=:w}.$$



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Matrix rank of RHS $-BB^{T} - U \operatorname{vec}^{-1}(w) U^{T}$ is $\leq r + 1!$ \rightsquigarrow Apply results for linear Lyapunov equations with r.h.s of rank r + 1.

= W



Theorem

[B./Breiten 2012]

Assume existence and uniqueness with stable A and $A_j = U_j V_j^T$, with $U_j, V_j \in \mathbb{R}^{n \times r_j}$. Set $r = \sum_{j=1}^m r_j$. Then the solution X of

$$AX + XA^{T} + \sum_{j=1}^{m} A_j XA_j^{T} + BB^{T} = 0$$

can be approximated by X_k of rank (2k+1)(m+r), with an error satisfying

$$||X - X_k||_2 \lesssim \exp(-\sqrt{k}).$$



- Generalized Alternating Directions Iteration (ADI) method.
 - 1. Computing square solution matrix ($\sim n^2$ unknowns) [DAMM 2008].
 - 2. Computing low-rank factors of solutions ($\sim n$ unknowns) [B./BREITEN 2013].
- Generalized Extended (or Rational) Krylov Subspace Method (EKSM/RKSM) [B./BREITEN 2013].
- Tensorized versions of standard Krylov subspace methods, e.g., PCG, PBiCGStab [Kressner/Tobler 2011, B./Breiten 2013].
- Inexact stationary (fix point) iteration [SHANK/SIMONCINI/SZYLD 2016].



Consider bilinear control systems:

$$\Sigma: \begin{cases} \dot{x}(t) = Ax(t) + \sum_{i=1}^{m} A_i x(t) u_i(t) + Bu(t), \\ y(t) = Cx(t), \quad x(0) = x_0, \end{cases}$$

where $A, A_i \in \mathbb{R}^{n \times n}, \ B \in \mathbb{R}^{n \times m}, \ C \in \mathbb{R}^{q \times n}$.



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Key Observation

[B./Breiten 2011]

Consider parameters as additional inputs, a linear parametric system

$$\dot{x}(t) = Ax(t) + \sum_{i=1}^{m_p} a_i(p)A_ix(t) + B_0u_0(t), \quad y(t) = Cx(t)$$

with $B_0 \in \mathbb{R}^{n \times m_0}$ can be interpreted as bilinear system: $u(t) := \begin{bmatrix} a_1(p) & \dots & a_{m_p}(p) & u_0(t) \end{bmatrix}^T$, $B := \begin{bmatrix} \mathbf{0} & \dots & \mathbf{0} & B_0 \end{bmatrix} \in \mathbb{R}^{n \times m}, \quad m = m_p + m_0$.



Linear parametric systems can be interpreted as bilinear systems.



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Consequence

Model order reduction techniques for bilinear systems can be applied to linear parametric systems!

Here: balanced truncation for bilinear systems.



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Alternative: \mathcal{H}_2 -optimal rational interpolation/bilinear IRKA [B./BREITEN 2012, B./BRUNS 2015, FLAGG/GUGERCIN 2015].



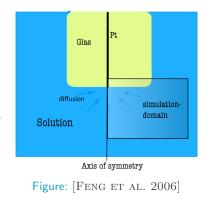
$$E\dot{x}(t) = (A + p_1(t)A_1 + p_2(t)A_2)x(t) + B,$$

$$y(t) = Cx(t), \quad x(0) = x_0 \neq 0,$$

- Rewrite as system with zero initial condition,
- FE model: n = 16,912, m = 3, q = 1,
- $p_j \in [0, 10^9]$ time-varying voltage functions,
- transfer function $G(i\omega, p_1, p_2)$,
- reduced system dimension r = 67,

•
$$\max_{\substack{\omega \in \{\omega_{min}, \dots, \omega_{max}\}\\ p_j \in \{p_{min}, \dots, p_{max}\}}} \frac{\|G - \hat{G}\|_2}{||G||_2} < 6 \cdot 10^{-4},$$

evaluation times: FOM 4.5h, ROM 38s
 → speed-up factor ≈ 426.



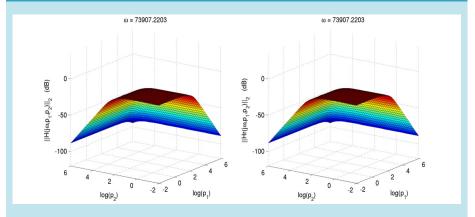
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Application to Parametric MOR

Fast Simulation of Cyclic Voltammogramms [FENG/KOZIOL/RUDNYI/KORVINK 2006]

Original...

and reduced-order model.





1. Introduction

- 2. Gramian-based Model Reduction for Linear Systems
- 3. Balanced Truncation for Bilinear Systems

4. Balanced Truncation for QB Systems

Balanced Truncation for Nonlinear Systems Gramians for QB Systems Truncated Gramians Numerical Results

5. Balanced Truncation for Polynomial Systems



• Nonlinear balancing based on energy functionals [SCHERPEN '93, GRAY/MESKO '96].

Definition

[Scherpen '93, Gray/Mesko '96]

The reachability energy functional, $L_c(x_0)$, and observability energy functional, $L_o(x_0)$ of a system are given as:

$$L_{c}(x_{0}) = \inf_{\substack{u \in L_{2}(-\infty,0] \\ x(-\infty)=0, \ x(0)=x_{0}}} \frac{1}{2} \int_{-\infty}^{0} \|u(t)\|^{2} dt, \qquad L_{o}(x_{0}) = \frac{1}{2} \int_{0}^{\infty} \|y(t)\|^{2} dt.$$

Disadvantage: energy functionals are the solutions of nonlinear Hamilton-Jacobi equations which are hardly solvable for large-scale systems.



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 Disadvantage: energy functionals are the solutions of nonlinear Hamilton-Jacobi equations which are hardly solvable for large-scale systems.
- Empirical Gramians/frequency-domain POD [Lall et al '99, Willcox/Peraire '02].

Example: controllability Gramian from time domain data (snapshots)

1. Define reachability Gramian of the system

 $P = \int_0^\infty x(t)x(t)^T dt$, where x(t) solves $\dot{x} = f(x, \delta)$, $x(0) = x_0$.

- 2. Use time-domain integrator to produce snapshots $x_k \approx x(t_k)$, $k = 1, \ldots, K$.
- 3. Approximate $P \approx \sum_{k=0}^{K} w_k x_k x_k^T$ with positive weights w_k .
- 4. Analogously for observability Gramian.
- 5. Compute balancing transformation and apply it to nonlinear system.

Disadvantage: Depends on chosen training input (e.g., $\delta(t_0)$) like other POD approaches.



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- ~ Goal: computationally efficient and input-independent method!

- W. S. Gray and J. P. Mesko. Controllability and observability functions for model reduction of nonlinear systems. In Proc. of the Conf. on Information Sci. and Sys., pp. 1244–1249, 1996.
- C. Himpe. emgr The empirical Gramian framework. ALGORITHMS 11(7): 91, 2018. doi:10.3390/a11070091.
- S. Lall, J. Marsden, and S. Glavaški. A subspace approach to balanced truncation for model reduction of nonlinear control systems. INTERNATIONAL JOURNAL OF ROBUST AND NONLINEAR CONTROL, 12:519-535, 2002.
- 🗎 J. M. A. Scherpen. Balancing for nonlinear systems. SYSTEMS & CONTROL LETTERS, 21:143–153, 1993.
- K. Willcox and J. Peraire, Balanced model reduction via the proper orthogonal decomposition. AIAA JOURNAL, 40:2323-2330, 2002.



- Nonlinear balancing based on energy functionals [SCHERPEN '93, GRAY/MESKO '96].
 Disadvantage: energy functionals are the solutions of nonlinear Hamilton-Jacobi equations which are hardly solvable for large-scale systems.
- Empirical Gramians/frequency-domain POD [LALL ET AL '99, WILLCOX/PERAIRE '02]. **Disadvantage:** Depends on chosen training input (e.g., $\delta(t_0)$) like other POD approaches.
- ~ Goal: computationally efficient and input-independent method!
- For recent developments on empirical Gramians, see [HIMPE '18].
- W. S. Gray and J. P. Mesko. Controllability and observability functions for model reduction of nonlinear systems. In Proc. of the Conf. on Information Sci. and Sys., pp. 1244–1249, 1996.
- 🖹 C. Himpe. emgr The empirical Gramian framework. ALGORITHMS 11(7): 91, 2018. doi:10.3390/a11070091.
- S. Lall, J. Marsden, and S. Glavaški. A subspace approach to balanced truncation for model reduction of nonlinear control systems. INTERNATIONAL JOURNAL OF ROBUST AND NONLINEAR CONTROL, 12:519-535, 2002.
- 🗎 J. M. A. Scherpen. Balancing for nonlinear systems. SYSTEMS & CONTROL LETTERS, 21:143–153, 1993.
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- A possible solution is to obtain bounds for the energy functionals, instead of computing them exactly.
- For bilinear systems, such local bounds were derived in [B./DAMM 2011] using the solutions to the Lyapunov-plus-positive equations:

$$AP + PA^{T} + \sum_{i=1}^{m} A_{i}PA_{i}^{T} + BB^{T} = 0,$$

$$A^{T}Q + QA^{T} + \sum_{i=1}^{m} A_{i}^{T}QA_{i} + C^{T}C = 0.$$

(If their solutions exist, they define reachability and observability Gramians of BIBO stable bilinear system.)

- Here we aim at determining algebraic Gramians for QB (and polynomial) systems, which
 - provide bounds for the energy functionals of QB systems,
 - generalize the Gramians of linear and bilinear systems, and
 - allow us to find the states that are hard to control as well as hard to observe in an efficient and reliable way.



- Gramians for QB Systems Controllability Gramians
- Consider input \rightarrow state map of QB system ($m = 1, N \equiv A_1$):

$$\dot{x}(t) = Ax(t) + Hx(t) \otimes x(t) + Nx(t)u(t) + Bu(t), \qquad x(0) = 0.$$

Integration yields

$$\begin{aligned} x(t) &= \int_{0}^{t} e^{A\sigma_{1}} Bu(t-\sigma_{1}) d\sigma_{1} + \int_{0}^{t} e^{A\sigma_{1}} Nx(t-\sigma_{1})u(t-\sigma_{1}) d\sigma_{1} \\ &+ \int_{0}^{t} e^{A\sigma_{1}} Hx(t-\sigma_{1}) \otimes x(t-\sigma_{1}) d\sigma_{1} \end{aligned}$$



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$$=\int_{0}^{t}e^{A\sigma_{1}}Bu(t-\sigma_{1})d\sigma_{1}+\int_{0}^{t}\int_{0}^{-1}e^{A\sigma_{1}}Ne^{A\sigma_{2}}Bu(t-\sigma_{1})u(t-\sigma_{1}-\sigma_{2})d\sigma_{1}d\sigma_{2}$$

$$+\int_{0}^{t}\int_{0}^{t-\sigma_{1}}\int_{0}^{t-\sigma_{1}}e^{A\sigma_{1}}H(e^{A\sigma_{2}}B\otimes e^{A\sigma_{3}}B)u(t-\sigma_{1}-\sigma_{2})u(t-\sigma_{1}-\sigma_{3})d\sigma_{1}d\sigma_{2}d\sigma_{3}+\ldots$$

[RUGH '81]



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$$= \int_{0}^{t} e^{A\sigma_{1}} Bu(t-\sigma_{1}) d\sigma_{1} + \int_{0}^{t} \int_{0}^{t-\sigma_{1}} e^{A\sigma_{1}} Ne^{A\sigma_{2}} Bu(t-\sigma_{1})u(t-\sigma_{1}-\sigma_{2}) d\sigma_{1} d\sigma_{2} \\ &+ \int_{0}^{t} \int_{0}^{t-\sigma_{1}} \int_{0}^{t-\sigma_{1}} e^{A\sigma_{1}} H(e^{A\sigma_{2}} B \otimes e^{A\sigma_{3}} B)u(t-\sigma_{1}-\sigma_{2})u(t-\sigma_{1}-\sigma_{3}) d\sigma_{1} d\sigma_{2} d\sigma_{3} + . \end{aligned}$$

 By iteratively inserting expressions for x(t − •), we obtain the Volterra series expansion for the QB system.



Using the Volterra kernels, we can define the controllability mappings

$$\begin{split} \Pi_1(t_1) &:= e^{At_1}B, \qquad \Pi_2(t_1, t_2) := e^{At_1}N\Pi_1(t_2), \\ \Pi_3(t_1, t_2, t_3) &:= e^{At_1}[H(\Pi_1(t_2) \otimes \Pi_1(t_3)), N\Pi_2(t_1, t_2)], \dots \end{split}$$

and a candidate for a new Gramian:

$$P := \sum_{k=1}^{\infty} P_k, \quad \text{where} \quad P_k = \int_0^{\infty} \cdots \int_0^{\infty} \Pi_k(t_1, \ldots, t_k) \Pi_k(t_1, \ldots, t_k)^T dt_1 \ldots dt_k.$$



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Theorem

[B./GOYAL '16]

If it exists, the new controllability Gramian P for QB (MIMO) systems with stable A solves the **quadratic Lyapunov equation**

$$AP + PA^{T} + \sum_{k=1}^{m} A_{k}PA_{k}^{T} + H(P \otimes P)H^{T} + BB^{T} = 0.$$

Note: $H = 0 \rightsquigarrow$ "bilinear reachability Gramian"; if additionally, all $A_k = 0 \rightsquigarrow$ linear one.



 Controllability energy functional (Gramian) of the dual system ⇔ observability energy functional (Gramian) of the original system.



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- Controllability energy functional (Gramian) of the dual system ⇔ observability energy functional (Gramian) of the original system.
- This allows to define dual systems for QB systems:

$$\begin{split} \dot{x}(t) &= Ax(t) + Hx(t) \otimes x(t) + \sum_{k=1}^{m} A_k x(t) u_k(t) + Bu(t), \qquad x(0) = 0, \\ \dot{x}_d(t) &= -A^T x_d(t) - H^{(2)} x(t) \otimes x_d(t) - \sum_{k=1}^{m} A_k^T x_d(t) u_k(t) - C^T u_d(t), \quad x_d(\infty) = 0, \\ y_d(t) &= B^T x_d(t), \end{split}$$

where $H^{(2)}$ is the mode-2 matricization of the QB Hessian.



• Writing down the Volterra series for the dual system \rightsquigarrow observability mapping.

• This provides the observability Gramian Q for the QB system. It solves

$$A^T Q + QA + \sum_{k=1}^m A_k^T QA_k + H^{(2)}(P \otimes Q) \left(H^{(2)}\right)^T + C^T C = 0.$$



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$$A^{\mathsf{T}}Q + QA + \sum_{k=1}^{m} A_k^{\mathsf{T}}QA_k + H^{(2)}(P \otimes Q) \left(H^{(2)}\right)^{\mathsf{T}} + C^{\mathsf{T}}C = 0.$$

Remarks:

- Observability Gramian depends on controllability Gramian!
- For H = 0, obtain "bilinear observability Gramian", and if also all $A_k = 0$, the linear one.



Bounding the energy functionals:

Lemma [B./GOYAL '17] In a neighborhood of the stable equilibrium, $B_{\varepsilon}(0)$, $L_{c}(x_{0}) \geq \frac{1}{2}x_{0}^{T}P^{-1}x_{0}, \qquad L_{o}(x_{0}) \leq \frac{1}{2}x_{0}^{T}Qx_{0}, \qquad x_{0} \in B_{\varepsilon}(0),$

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for "small signals" and x_0 pointing in unit directions.

Another interpretation of Gramians in terms of energy functionals

- 1. If the system is to be steered from 0 to x_0 , where $x_0 \notin \operatorname{range}(P)$, then $L_c(x_0) = \infty$ for all feasible input functions u.
- 2. If the system is (locally) controllable and $x_0 \in \ker(Q)$, then $L_o(x_0) = 0$.



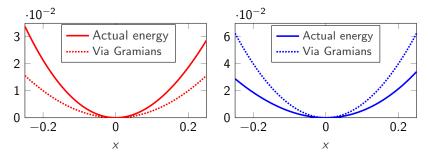
Illustration using a scalar system

$$\dot{x}(t) = ax(t) + hx^2(t) + nx(t)u(t) + bu(t), \quad y(t) = cx(t).$$



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(a) Input energy lower bound.
(b) Output energy upper bound.
Figure: Comparison of energy functionals for -a = b = c = 2, h = 1, n = 0.



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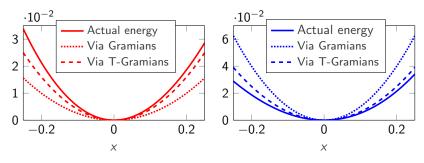
[DAMM '08]

• To overcome this issue, we propose truncated Gramians for QB systems.

Definition (Truncated Gramians) [B./GOVAL '16]
The truncated Gramians
$$P_{\mathcal{T}}$$
 and $Q_{\mathcal{T}}$ for QB systems satisfy
 $AP_{\mathcal{T}} + P_{\mathcal{T}}A^T = -BB^T - \sum_{k=1}^m A_k P_l A_k^T - H(P_l \otimes P_l) H^T$,
 $A^T Q_{\mathcal{T}} + Q_{\mathcal{T}}A = -C^T C - \sum_{k=1}^m A_k^T Q_l A_k - H^{(2)}(P_l \otimes Q_l)(H^{(2)})^T$,
where
 $AP_l + P_l A^T = -BB^T$ and $A^T Q_l + Q_l A = -C^T C$.







(a) Input energy lower bounds. (b) Output energy upper bounds.

Figure: Comparison of energy functionals for -a = b = c = 2, h = 1, n = 0.



• $\sigma_i(P \cdot Q) > \sigma_i(P_T \cdot Q_T) \Rightarrow$ obtain smaller order of reduced system if truncation is done at the same cutoff threshold.



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- Most importantly, we need solutions of only four standard Lyapunov equations.



- T-Gramians approximate energy functionals better than the actual Gramians.
- $\sigma_i(P \cdot Q) > \sigma_i(P_T \cdot Q_T) \Rightarrow$ obtain smaller order of reduced system if truncation is done at the same cutoff threshold.
- Most importantly, we need solutions of only four standard Lyapunov equations.
- Interpretation of controllability/observability of the system via T-Gramians:
 - If the system is to be steered from 0 to x_0 , where $x_0 \notin range(P_T)$, then $L_c(x_0) = \infty$.
 - If the system is controllable and $x_0 \in \ker(Q_T)$, then $L_o(x_0) = 0$.



1: **Input:** A, H, A_k, B, C .



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5: Output: reduced-order matrices:

$$\hat{A} = \mathcal{W}^{\mathsf{T}} A \mathcal{V}, \quad \hat{H} = \mathcal{W}^{\mathsf{T}} H(\mathcal{V} \otimes \mathcal{V}), \quad \hat{A}_{k} = \mathcal{W}^{\mathsf{T}} A_{k} \mathcal{V}, \\ \hat{B} = \mathcal{W}^{\mathsf{T}} B, \quad \hat{C} = C \mathcal{V}.$$

Remark: There are efficient ways to compute \hat{H} , avoiding the explicit computation of $\mathcal{V} \otimes \mathcal{V}$. [B./BREITEN '15, B./GOYAL/GUGERCIN '18]



$$\begin{aligned} v_t + v^3 &= v_{xx} + v, & (0, L) \times (0, T), \\ v(0, .) &= u(t), & (0, T), \\ v_x(L, .) &= 0, & (0, T), \\ v(x, 0) &= v_0(x), & (0, L). \end{aligned}$$

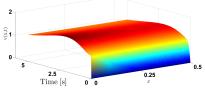
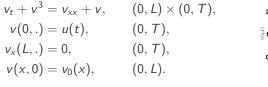


Figure: Chafee-Infante equation.

• Cubic nonlinearity that can be rewritten into QB form. [B./BREITEN '15']





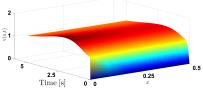
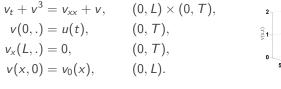


Figure: Chafee-Infante equation.

- Cubic nonlinearity that can be rewritten into QB form. [B./BREITEN '15']
- The transformed QB system is of order n = 1,000.
- The output of interest is the response at right boundary at x = L.





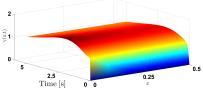


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- Cubic nonlinearity that can be rewritten into QB form. [B./BREITEN '15']
- The transformed QB system is of order n = 1,000.
- The output of interest is the response at right boundary at x = L.
- We determine the reduced-order system of order r = 10.



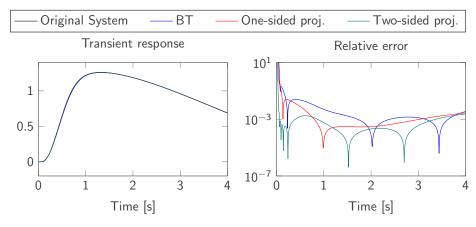


Figure: Boundary control for a control input $u(t) = 5t \exp(-t)$.



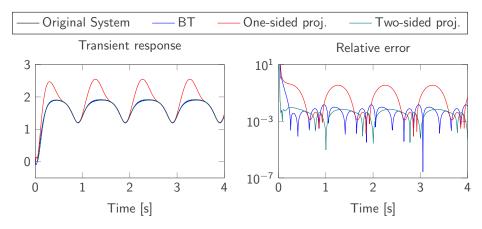


Figure: Boundary control for a control input $u(t) = 25(1 + \sin(2\pi t))/2$.



$$\begin{aligned} \epsilon v_t(x,t) &= \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + q, \\ w_t(x,t) &= hv(x,t) - \gamma w(x,t) + q, \end{aligned}$$

Numerical Results

FitzHugh-Nagumo (F-N) model

with a nonlinear function

$$f(v(x, t)) = v(v - 0.1)(1 - v).$$

The boundary conditions are as follows:

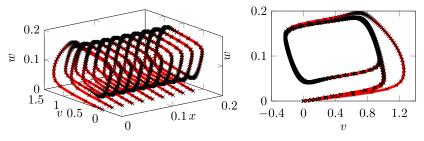
$$v_x(0,t) = i_0(t), \quad v_x(L,t) = 0, \quad t \ge 0,$$

where $\epsilon = 0.015$, h = 0.5, $\gamma = 2$, q = 0.05, L = 0.2.

• Input $i_0(t) = 5 \cdot 10^4 t^3 \exp(-15t)$ serves as actuator.



Original system (n = 1500) × Reduced system (BT) (r = 20)



(a) Limit-cycles at various x. (b) Projection onto the v-w plane.

Figure: Comparison of the limit-cycles obtained via the original and reduced-order (BT) systems. The reduced-order systems constructed by moment-matching methods were unstable.



1. Introduction

- 2. Gramian-based Model Reduction for Linear Systems
- 3. Balanced Truncation for Bilinear Systems
- 4. Balanced Truncation for QB Systems
- 5. Balanced Truncation for Polynomial Systems Polynomial Control Systems Gramians for PC Systems Truncated Gramians Numerical Example



Now, consider the class of polynomial control (PC) Systems:

$$\begin{split} \dot{x}(t) &= Ax(t) + \sum_{j=2}^{n_p} H_j\left(\otimes^j x(t)\right) + \sum_{j=2}^{n_p} \sum_{k=1}^m N_j^k\left(\otimes^j x(t)\right) u_k(t) + Bu(t),\\ y(t) &= Cx(t), \ x(0) = 0, \end{split}$$

where

• n_p is the degree of the polynomial part of the system,

•
$$x(t) \in \mathbb{R}^n$$
, $\otimes^j x(t) = \underbrace{x(t) \otimes \cdots \otimes x(t)}_{i \in \mathbb{N}}$,

j-times

•
$$u(t) \in \mathbb{R}^m$$
, and $y(t) \in \mathbb{R}^p$, $n \gg m, p$.

- $A \in \mathbb{R}^{n \times n}$, $H_j, N_j^k \in \mathbb{R}^{n \times n^j}$, $B \in \mathbb{R}^{n \times m}$ and $C \in \mathbb{R}^{p \times n}$.
- Assumption: A is supposed to be Hurwitz \Rightarrow local stability.



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- Assumption: A is supposed to be Hurwitz \Rightarrow local stability.

Examples: FitzHugh-Nagumo and Chafee-Infante equations lead to cubic control systems; cubic-quintic Allen-Cahn equation to quintic control system.



Expanding the response of the PC system into a Volterra series representation and following the same ideas as in the QB case, we define the reachability Gramian as

$$P = \sum_{k=1}^{\infty} \int_0^{\infty} \cdots \int_0^{\infty} \bar{P}_k(t_1, \ldots, t_k) \bar{P}_k(t_1, \ldots, t_k)^T dt_1 \ldots dt_k,$$



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where $\bar{P}_1(t_1) = e^{At_1}B$, $\bar{P}_2(t_1, t_2) = \sum_{k=1}^{m} e^{At_1}N_1^k e^{At_2}B$, $\bar{P}_3(t_1, t_2, t_3) = e^{At_1}H_2e^{At_2}B \otimes e^{At_3}B$,... are the kernels of the Volterra series.



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 $\bar{P}_3(t_1, t_2, t_3) = e^{At_1}H_2e^{At_2}B \otimes e^{At_3}B, \dots$ are the kernels of the Volterra series.

m

Theorem

The reachability Gramian \mathbf{P} of a PC system solves the polynomial Lyapunov equation

$$AP + PA^{T} + BB^{T} + \sum_{j=2}^{n_{p}} H_{j}\left(\otimes^{j} P\right) H_{j}^{T} + \sum_{j=2}^{n_{p}} \sum_{k=1}^{m} N_{j}^{k}\left(\otimes^{j} P\right)\left(N_{j}^{k}\right)^{T} = 0.$$



The Observability Gramian is defined as follows



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• First, we write the adjoint system as

[Fujimoto et. al. '02]

$$\begin{split} \dot{x}(t) &= Ax(t) + \sum_{j=2}^{n_p} H_j x_j^{\otimes}(t) + \sum_{j=1}^{n_p} \sum_{k=1}^m N_j^k x_j^{\otimes}(t) u_k(t) + Bu(t), \\ \dot{x}_d(t) &= -A^T x_d(t) - \sum_{j=2}^{n_p} H_j^{(2)} x_{d,j}^{\otimes}(t) - \sum_{j=1}^{n_p} \sum_{k=1}^m \left(N_j^{k,(2)}\right) x_{d,j}^{\otimes}(t) u_{d,k}(t) - C^T u_d(t), \quad x_d(\infty) = 0, \\ y_d(t) &= B^T x_d(t). \end{split}$$



The Observability Gramian is defined as follows

• First, we write the adjoint system as

[Fujimoto et. al. '02]

$$\begin{split} \dot{\boldsymbol{x}}(t) &= \boldsymbol{A}\boldsymbol{x}(t) + \sum_{j=2}^{n_p} \boldsymbol{H}_j \boldsymbol{x}_j^{\otimes}(t) + \sum_{j=1}^{n_p} \sum_{k=1}^m N_j^k \boldsymbol{x}_j^{\otimes}(t) \boldsymbol{u}_k(t) + \boldsymbol{B}\boldsymbol{u}(t), \\ \dot{\boldsymbol{x}}_d(t) &= -\boldsymbol{A}^T \boldsymbol{x}_d(t) - \sum_{j=2}^{n_p} \boldsymbol{H}_j^{(2)} \boldsymbol{x}_{d,j}^{\otimes}(t) - \sum_{j=1}^{n_p} \sum_{k=1}^m \left(N_j^{k,(2)}\right) \boldsymbol{x}_{d,j}^{\otimes}(t) \boldsymbol{u}_{d,k}(t) - \boldsymbol{C}^T \boldsymbol{u}_d(t), \quad \boldsymbol{x}_d(\infty) = \boldsymbol{0}, \\ \boldsymbol{y}_d(t) &= \boldsymbol{B}^T \boldsymbol{x}_d(t). \end{split}$$

• Then, by taking the kernel of Volterra series, one has

Theorem

Let ${\bf P}$ be the reachability Gramian. Then, the observability Gramian ${\bf Q}$ of a PC system solves the polynomial Lyapunov equation

$$A^{T}Q + QA + C^{T}C + \sum_{j=2}^{n_{p}} H_{j}^{(2)} \left(\otimes^{j-1}P \otimes Q \right) \left(H_{j}^{(2)} \right)^{T} + \sum_{j=2}^{n_{p}} \sum_{k=1}^{m} N_{j}^{k,(2)} \left(\otimes^{j-1}P \otimes Q \right) \left(N_{j}^{k,(2)} \right)^{T} = 0.$$



- Polynomial Lyapunov equations are very expensive to solve.
- As for QB systems, we thus propose truncated Gramians that only involve a finite number of kernels.

$$P_{\mathcal{T}} = \sum_{k=1}^{n_p+1} \int_0^\infty \cdots \int_0^\infty \bar{P}_k(t_1,\ldots,t_k) \bar{P}_k(t_1,\ldots,t_k)^T dt_1 \ldots dt_k,$$

Truncated Gramians

whe

The reachability truncated Gramian solves

$$AP_{\mathcal{T}} + P_{\mathcal{T}}A^{T} + BB^{T} + \sum_{j=2}^{n_{p}} H_{j} \otimes^{j} P_{l}H_{j}^{T} + \sum_{j=2}^{n_{p}} \sum_{k=1}^{m} N_{j}^{k} \otimes^{j} P_{l} \left(N_{j}^{k}\right)^{T} = 0.$$

we $AP_{l} + P_{l}A^{T} + BB^{T} = 0$

• Advantage: Only need to solve a finite number of (linear) Lyapunov equations.



$$\epsilon v_t(x,t) = \epsilon^2 v_{xx}(x,t) + f(v(x,t)) - w(x,t) + q,$$

$$w_t(x,t) = hv(x,t) - \gamma w(x,t) + q,$$

with a nonlinear function

f(v(x, t)) = v(v - 0.1)(1 - v).

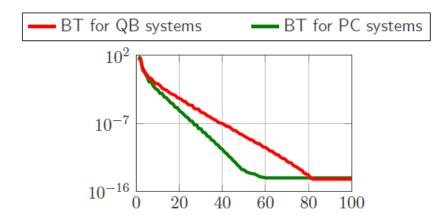
The boundary conditions are as follows:

 $v_x(0,t) = i_0(t), \quad v_x(L,t) = 0, \quad t \ge 0,$

where $\epsilon = 0.015, \ h = 0.5, \ \gamma = 2, \ q = 0.05, \ L = 0.2.$

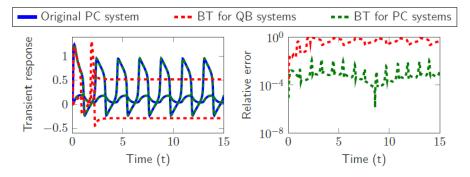
- After discretization we obtain a PC system with cubic nonlinearity of order $n_{pc} = 600.$ [B./BREITEN '15]
- The transformed quadratic-bilinear (QB) system is of order $n_{qb} = 900$.
- The outputs of interest v(0, t), w(0, t) are the responses at the left boundary at x = 0.
- We compare balanced truncation for PC and QB systems.





• Decay singular values for PC systems is faster \Rightarrow smaller reduced order model!





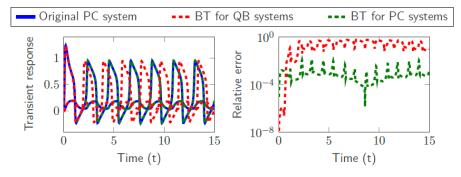
Numerical Example

Time-domain simulations

• Original PC system of order 600. Original QB system of order 900.

• Reduced PC system of order 10. Reduced QB system of order 10.





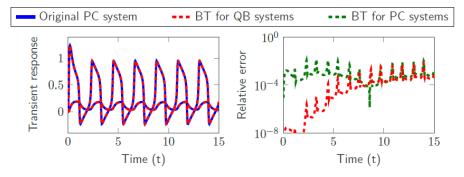
Numerical Example

Time-domain simulations

• Original PC system of order 600. Original QB system of order 900.

• Reduced PC system of order 10. Reduced QB system of order 30.





• Original PC system of order 600. Original QB system of order 900.

• Reduced PC system of order 10. Reduced QB system of order 43.



- BT extended to bilinear, QB, and polynomial systems.
- Local Lyapunov stability is preserved.
- As of yet, only weak motivation by bounding energy functionals.
- No error bounds in terms of "Hankel" singular values.
- Computationally efficient (as compared to nonlinear balancing), and input independent.

• To do:

- improve efficiency of Lyapunov solvers with many right-hand sides further;
- error bound;
- conditions for existence of new QB Gramians;
- extension to descriptor systems;
- time-limited versions.



P. Benner and T. Damm.

Lyapunov Equations, Energy Functionals, and Model Order Reduction of Bilinear and Stochastic Systems. SIAM JOURNAL ON CONTROL AND OPTIMIZTION, 49(2):686–711, 2011.

- P. Benner and T. Breiten. Low Rank Methods for a Class of Generalized Lyapunov Equations and Related Issues. NUMERISCHE MATHEMATIK, 124(3):441–470, 2013.
- T. Damm and P. Benner.

Balanced Truncation for Stochastic Linear Systems with Guaranteed Error Bound. In PROCEEDINGS OF MTNS 2014, pp. 1492–1497, 2014.

- P. Benner, T. Damm, M. Redmann, and Y. Rocio Rodriguez Cruz. Positive Operators and Stable Truncation. LINEAR ALGEBRA AND ITS APPLICATIONS, 498:74–87, 2016.
- P. Benner, T. Damm, M. Redmann, and Y. Rocio Rodriguez Cruz. Dual Pairs of Generalized Lyapunov Inequalities and Balanced Truncation of Stochastic Linear Systems. IEEE TRANSACTIONS on AUTOMATIC CONTROL, 62(2):782–791, 2017.

P. Benner, P. Goyal, and M. Redmann.

Truncated Gramians for Bilinear Systems and their Advantages in Model Order Reduction. In P. Benner, M. Ohlberger, T. Patera, G. Rozza, K. Urban (Eds.), MODEL REDUCTION OF PARAMETRIZED SYSTEMS, MS & A — Modeling, Simulation and Applications, Vol. 17, pp. 285–300. Springer International Publishing, Cham, 2017.

P. Benner and P. Goyal.

Balanced Truncation Model Order Reduction for Quadratic-Bilinear Control Systems. arXiv Preprint arXiv:1705.00160, April 2017.

P. Benner, P. Goyal, and I. Pontes Duff. Approximate Balanced Truncation for Polynomial Control Systems.

In preparation.